

Digital PID Control System for DC Servo Motor Using VHDL Code

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Abstract: This project investigates based PID motion control systems for small, self- adaptive systems. The closed loop position control of DC servo motor is performed using PWM signal. VHDL based PID motion control system provides an efficient and cheap method. Proportion (P)–increases gain margin, increase system response speed. Integration (I)–minimizes steady state error Differentiation (D)–increases system stability. VHDL: HARDWARE DESCRIPTION LANGUAGES (HDL'S) are used to describe hardware for the purpose of simulation, modeling, testing design and documentation of digital systems. PWM PULSE WIDTH MODULATION which circuit works by making a square wave. Advantage of pulse width modulation is that the pulses reach the full supply voltage and will produce more torque in a motor

Keywords: component; formatting; style; styling; insert (key words)

1. Introduction

The PID controller provides,

- P – Increases gain margin, increase system response speed
- I – Minimizes steady state error
- D – Increases system stability

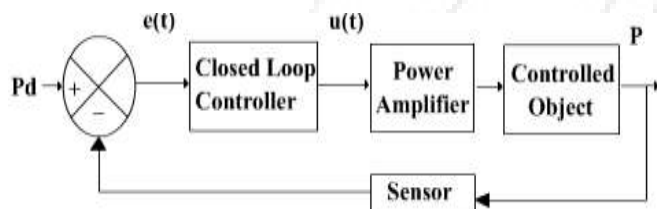
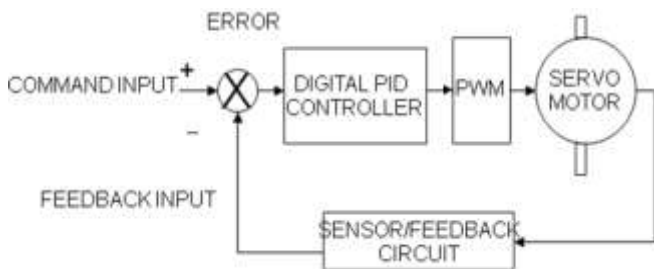


Figure: Closed loop control system

A closed loop control system is shown in figure which is used to control a device such as a servo motor. P and P_d correspond to the controlled variable (e.g. rotational position) and its desired value, which is provided at a higher control level. The goal is to eliminate the error between P and P_d. The value of P is measured by the sensor, which is compared with P_d to generate the error e(t). The output to the controlled device, u(t), from the closed-loop controller is a function of e(t). Typically this is a weak signal that requires amplification.

2. Block Diagram of PID Control System



PID Control Algorithm

- A closed-loop control system is used to control a device such as a servo motor .

- P and P_d correspond to the controlled variable (e.g. rotational position).
- P_d Goal is to eliminate the error between P and P_d.
- Value of P is measured by the sensor, which is compared with P_d to generate the error e(t).
- The output to the controlled device u(t).
- Closed-loop controller is e(t).

$$u(t) = kp[e(n) + 1/Ti \int_0^t e(t)dt + Td * de(t) / dt]$$

- For a small sample interval T, this above equation can be turned into a difference equation by discretization .

$$u(n) = kp(e(n)) + Ki \sum_{j=0}^n e(j) + Kd(e(n) - e(n-1))$$

- A difference equation can be implemented by a digital system, either in hardware or software.
- The derivative term is simply replaced by a first-order difference expression and the integral by a sum, thus the **difference equation** is given as:

PID Control Algorithm

$$u(t) = kp[e(n) + 1/Ti \int_0^t e(t)dt + Td * de(t) / dt]$$

u(t)=The output to the controlled device

Kp = proportional gain

Kd = derivative gain

e = error

KI = integral gain

Ti-integral time constant,

Td-derivative time constant

$$u(n) = kp(e(n)) + KpT / Ti \sum_{j=0}^n e(j) + KpTd / T(e(n) - e(n-1))$$

$$u(n) = kp(e(n)) + Ki \sum_{j=0}^n e(j) + Kd(e(n) - e(n-1))$$

$$u(n-1) = kp(e(n-1)) + Ki \sum_{j=0}^{n-1} e(j) + Kd(e(n-1) - e(n-2))$$

$$\Delta u(n) = u(n) - u(n-1)$$

$$\Delta u(n) = kp(e(n)) + Ki \sum_{j=0}^n e(j) + Kd(e(n) - e(n-1)) - [kp(e(n-1)) + Ki \sum_{j=0}^{n-1} e(j) + Kd(e(n-1) - e(n-2))]$$

$$\Delta u(n) = kp(e(n) - e(n-1)) + Ki(\sum_{j=0}^n e(j) - \sum_{j=0}^{n-1} e(j)) + kd(e(n) - e(n-1) - ((e(n) - e(n-2))))$$

$$\Delta u(n) = kp(e(n) - e(n-1)) + kd(e(n) - e(n-1) - e(n-1) + e(n-2)) + Ki[e(0) + e(1) + \dots + e(n)] - (e(0) + e(1) + \dots + e(n-1))$$

$$\Delta u(n) = Kp(e(n) - e(n-1)) + kd(e(n) - 2e(n-1) + e(n-2)) + Ki(e(n))$$

$$\Delta u(n) = Kp(e(n)) - Kpe(n-1) + kd(e(n)) - Kd2e(n-1) + Kde(n-2) + Ki(e(n))$$

$$u(t) = kp[e(n) + 1/Ti \int_0^t e(t) dt + Td * de(t) / dt]$$

$$\Delta u(n) = (Kp + Ki + kd)e(n) + (-Kp - 2Kd)e(n-1) + Kde(n-2)$$

$$\Delta u(n) = K_0(e(n) + K_1(e(n-1) + K_2e(n-2)))$$

step input

$$K_0 = (Kp + Ki + kd)$$

$$K_1 = (-Kp - 2Kd)$$

$$K_2 = Kd$$

$$\Delta u(n) = u(n) - u(n-1) \dots \dots \dots \text{equation (1)}$$

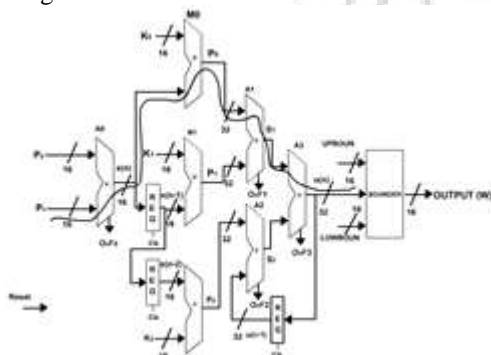
from equation (1) we find

$$u(n) = \Delta u(n) + u(n-1)$$

$$u(n) = u(n-1) + K_0(e(n) + K_1(e(n-1) + K_2(n-2)))$$

PID DESIGN

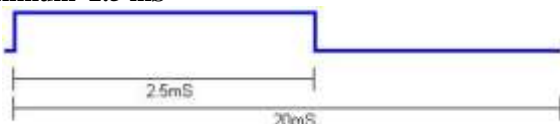
- The design requires 4 adders and 3 multipliers.
- Signal clk is used to control sampling frequency.
- Encoder counter value, represents the current position P.
- The negation of P, P neg, is generated by bit-wise complementing and adding 1.
- Latched at register REG0, thus becomes e(n-1)
- e(n-2) and u(n-1) are recorded at REG 1 and REG 2 by latching (n-1) and u(n) respectively.
- p0, p1, p2, s1, s2 are temporary variables.
- $e(n) = pd + (-p)$
- $p0 = k0 * e(n)$
- $p1 = k1 * e(n-1)$
- $p2 = k2 * e(n-2)$
- $s1 = p0 + p1$
- $s2 = p2 + u(n-1)$
- $u(n) = s1 + s2$
- PID design



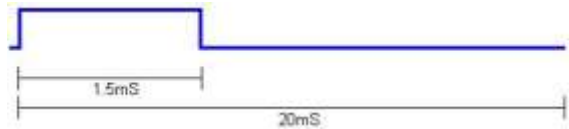
Operation of Servomotor

A typical servo consists of a DC motor, a gear head, a potentiometer for position feedback, and a small circuit to read the pot and position the motor. Position Pulse width

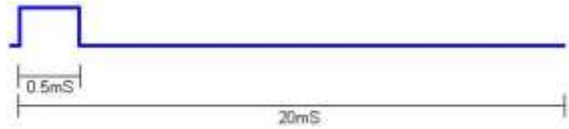
Maximum 2.5 mS



Centre 1.5mS

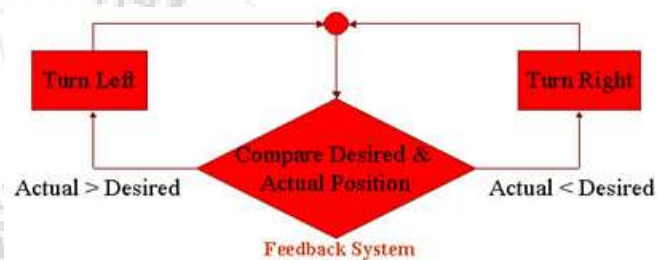


Minimum 0.5mS



For example, sending a 1.5 ms pulse to the servo, tells the servo that the desired position is 90 degrees. In order for the servo to hold this position, the command must be sent at about 50Hz or every 20ms.

Once the servo has received the desired position (via the PWM signal) the servo must attempt to match the desired and actual positions. It does this by turning a small, geared motor left or right. If, for example, the desired position is less than the actual position, the servo will turn to the left. On the other hand, if the desired position is greater than the actual position, the servo will turn to the right



Specifications of Servo Motor

- Voltage: 24V DC
- Current : 1.0 Amp
- Speed: 1500 rpm
- Armature Inductance: 2.2mH
- Torque: 1.50 N cm
- Torque constant: 12.44 N cm/A Moment of Inertia : 0.25 Kg cm²

General Equations of Motor

- $V = IR + L \frac{di}{dt} + E$
- $T = j \frac{dw}{dt} + Bw + Tf$
- where, V= voltage in volts,
- T= Torque in N cm, ØØØØ
- I= current in Amperes, J= Moment of Inertia in Kg-cm²
- R= Resistance in ohms, B= Viscous coefficient of friction in Nm-s/rad,
- L= Inductance in henry, Tf= Load Torque in N cm,
- w=angular velocity in rad/s
- The position transfer function obtained by taking laplace transform of above equation is

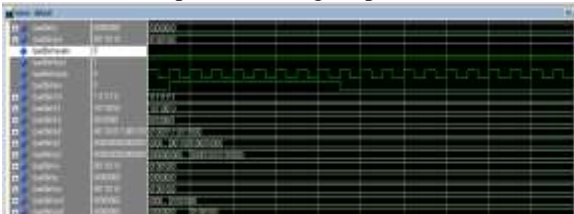
$$(\Theta)/v(s) = 6.374/s(12.421s+1)$$

Simulation Output

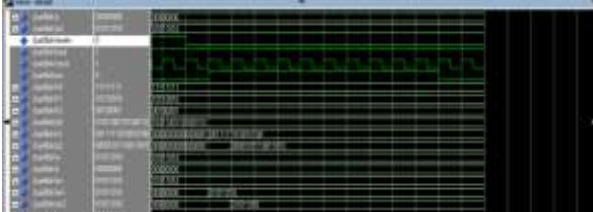
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x 1111111111111111
y 111111111101100
pd 00000000000101000
ub 0111111111111111
lb 0000000000000000
k0 0000000000001001
k1 0000000000000111
k2 0000000000000010
resetn 0
clock 1
cin0 0
cin1 0
cin2 0
cin3 0
un1 0000000000000000000110010101000
w 0000110010101000
cout0 0
cov0 0
ov0 0
ov1 0
ov2 0
ov3 0
p0 000000000000000000001000011100
p1 000000000000000000000110100100
p2 00000000000000000000000111000
s1 0000000000000000000001111000000
s2 0000000000000000000011010010000
un 000000000000000000000100011100000
p 00000000000111100
en 0000000000111100
en1 000000000111100
en2 000000000111100
v 0000110010101000
    
```

Servo motor output for 72 degree position



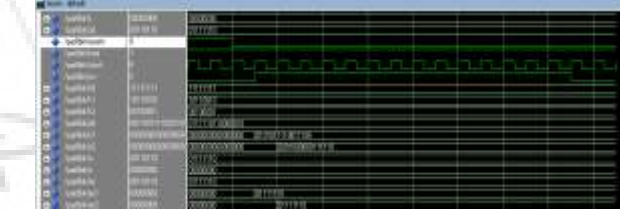
Servo motor output for 90 degree position



Servo motor output 108 degree position



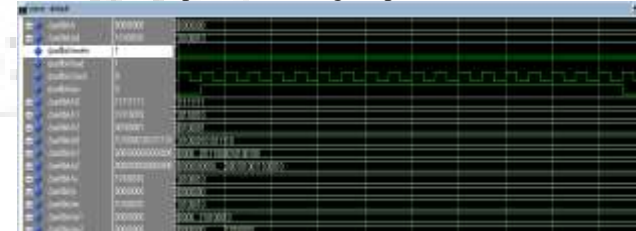
Servo motor output for 126 degree position



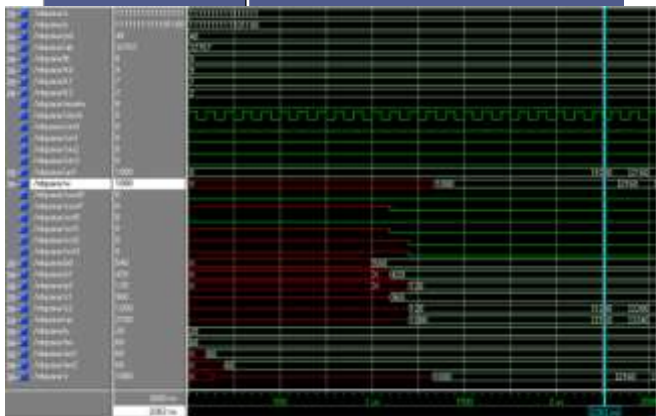
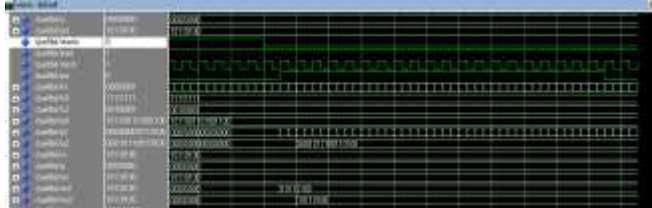
Servo motor output for 144 degree position



Servo motor output for 162 degree position

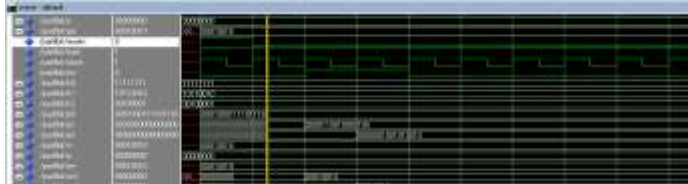


Servo motor output for 180 degree position

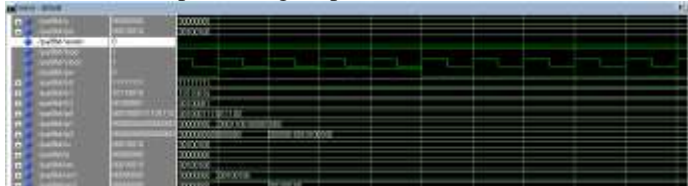


Degree Position	Binary value	Clock pluses
0	00000000	0
18	00010010	02
36	00100100	04
54	00110110	06
72	01001000	08
90	01011010	10
108	01101100	12
126	01111110	14
144	10010000	16
162	10100010	18
180	10110100	20

Servo Motor for 18 degree position



Servo motor output 36 degree position



Servomotor output for 54 degree position



References

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