An Improved Efficiency of Integrated Inverter / Converter for Dual Mode EV/HEV Application

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Abstract: The objective of this paper is to improve the efficiency of EV/HEV under heavy load conditions. The proposed integrated circuit allows the permanent magnet synchronous motor to operate in motor mode or acts as boost inductors of the boost converter and thereby boosting the output torque coupled to the same transmission system or dc-link voltage of the inverter connected to the output of the integrated circuit. In motor mode, the proposed integrated circuit acts as an inverter and it becomes a boost-type boost converter, while using the motor windings as the boost inductors to boost the converter output voltage. The proposed control technique is to use interleaved control to significantly reduce the current ripple and thereby reducing the losses and thermal stress under heavy-load condition. The performance of the proposed circuit is available by using simulation of MATLAB/SIMULINK software

Keywords: Hybrid Electric Vehicles (HEV), Boost Converter, Electric Vehicles (EV), Internal Combustion Engine (ICE)

1. Introduction

IN PARALLEL hybrid electric vehicle (HEV) [1]–[3] and electric vehicle (EV) [4], [5] system as shown in Fig. 1, the converter is used for boosting the battery voltage to rated dc bus for an inverter to drive motor. In the multimotor drive system [6], [7], the system will use two or more motors to boost torque, especially under low speed and high-torque region as shown in Fig. 2. For such applications, two or more inverters/converters are required. Fig. 3 shows the application of the proposed integrated circuit for motor drives with dual-mode control for EV/HEV applications. As shown in Fig. 3, the proposed integrated circuit allows the permanent magnet synchronous motor (PMSM) to operate in motor mode or acts as boost inductors of the boost converter and thereby boosting the output torque coupled to the same transmission system or dc-link voltage of an inverter connected to the output of the integrated circuit. In motor mode, the proposed integrated circuit acts as an inverter and it becomes a boost-type boost converter, while using the motor windings as the boost inductors to boost the converter output voltage. Therefore, the proposed integrated circuit can significantly reduce the volume and weight of the system.

Figure 1: HEV and EV system (a) Parallel HEV drive train. (b) EV drive train.

Figure 2: Conventional multi motor drive system of EV/HEV

Figure 3: Proposed integrated inverter/converter for the multi motor drive system
The integrated circuit presented in this paper can act as an inverter and a boost converter depending on the operation mode. For the integrated circuit, it not only can reduce the volume and weight but also boost torque and dc-link voltage for motor/converter modes, respectively. Moreover, a new control technique for the proposed integrated circuit under boost converter mode is proposed to increase the efficiency. For conventional circuit, shown in Fig. 4, a single phase boost converter [8] has been widely used for boost control due to its simplicity. However, for higher power applications, an interleaved boost converter [9]–[13] can reduce the current ripple and components stress and thereby reducing the losses and thermal stress. Based upon the interleaved control idea, a boost-control technique using motor windings as boost inductors for the proposed integrated circuit will be proposed. Under light load, the integrated circuit acts as a single-phase boost converter for not invoking additional switching and conduction losses, and functions as the two-phase interleaved boost converter under heavy load to significantly reduce the current ripple and thereby reducing the losses and thermal stress. Therefore, the proposed control technique for the proposed integrated circuit under boost converter mode can increase the efficiency.

When the proposed integrated circuit is operated in the converter mode, relay is turned OFF. And a single-phase or inter leaved control method will be applied to control of the power devices depending upon the load conditions.

B. Modelling and Controller Design Under Boost Mode

This section will introduce the model of boost converter and derive the transfer function of the voltage controller. Fig. 7 shows the nonideal equivalent circuit of the boost converter, it considers nonideal condition of components: inductor winding resistance $R_L$, collector-emitter saturation voltage $V_{CE}$, diode forward voltage drop $V_D$, and equivalent series resistance of capacitor $R_{esr}$. Analysis of the boost converter by using the state-space averaging method [14], small-signal ac equivalent circuit can be derived, as shown in Fig. 8.

Table 1: Specifications of Components

<table>
<thead>
<tr>
<th>Component</th>
<th>Specification</th>
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<tbody>
<tr>
<td>$C_{in}$/$C_{out}$</td>
<td>330uF/450V</td>
</tr>
<tr>
<td>Diode</td>
<td>$V_R=500V$, $I_P=20A$, $V_F=0.87V$</td>
</tr>
<tr>
<td>IGBT</td>
<td>$V_{CE}=600V$, $I_C=75A$, $V_{CE(sat)}=1.5V$</td>
</tr>
</tbody>
</table>

Fig. 5 shows the integrated circuit for dual-mode control. In Fig. 5, $C_{in}$ and $C_{out}$ can stabilize the voltage when input and output voltages are disturbed by source and load, respectively. Diode ($D$) is used for preventing output voltage impact on the input side. When the integrated circuit is operated in inverter (motor) mode, relay will be turned ON and six power devices (IGBTs in Fig. 5) are controlled by pulse width modulation (PWM) control signals. Details of the component specifications are shown in Table I.
By Fig. 8, the transfer function of the voltage controller can be derived as shown in (1).

\[
G_{vd}(s) = \frac{-6.787 \times 10^5 s^2 + 0.6837 s + 2498}{2.034 \times 10^{-5} s^2 + 0.00409 s + 3.242} \quad (2)
\]

Fig. 9 shows the block diagram of voltage loop, using a proportional-integral (PI) controller for the compensator. In this paper, the switching frequency is 20 kHz and voltage loop bandwidth will be less than 2 kHz. And the phase margin should be more than 45° to enhance the noise immunity. For the designed controller shown in (3), the Bode plot of the closed loop gain as shown in Fig. 10, the bandwidth is 7.73 Hz and the phase margin is 91.8°.

\[
C(s) = \frac{0.024897 s + 13.078}{s^2 L C (R + R_{ser}) + (L + C L) R_{ser} + (1 - D)(1 - L) R_{L} + (1 - D) R_{L} + R_L} \quad (3)
\]

### Table 2: Controller Design Parameters

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Value</th>
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<tbody>
<tr>
<td>(V_{in})</td>
<td>96 VDC</td>
</tr>
<tr>
<td>(V_o)</td>
<td>288 VDC</td>
</tr>
<tr>
<td>(P_o)</td>
<td>3 kW</td>
</tr>
<tr>
<td>(C)</td>
<td>260 (\mu)F</td>
</tr>
<tr>
<td>(R_L)</td>
<td>170 mΩ</td>
</tr>
<tr>
<td>(L)</td>
<td>2.77 mH</td>
</tr>
<tr>
<td>(R_{ser})</td>
<td>108 mΩ</td>
</tr>
<tr>
<td>(V_{d})</td>
<td>0.462 V</td>
</tr>
<tr>
<td>(V_{CE})</td>
<td>1.5 V</td>
</tr>
<tr>
<td>Voltage drop of (D)</td>
<td>0.87 V</td>
</tr>
</tbody>
</table>

Substituting the parameters shown in Table II into (1) give

\[
G_{vd}(s) = \frac{-6.787 \times 10^5 s^2 + 0.6837 s + 2498}{2.034 \times 10^{-5} s^2 + 0.00409 s + 3.242} \quad (2)
\]

3. Simulation Results

Here the simulation carried by two different cases they are 1) Proposed interleaved boost converter multiplier module 2) PV as input source of proposed converter with interleaved boost converter

Case-1 Proposed interleaved boost converter

\[
G_{vd}(s) = \frac{-6.787 \times 10^5 s^2 + 0.6837 s + 2498}{2.034 \times 10^{-5} s^2 + 0.00409 s + 3.242} \quad (2)
\]
Figure 11.1: measured current with and without interleaved control, Single-phase interleaved boost converter.

Figure 11.2: Measured current with and without interleaved control, Two-phase interleaved boost converter.

Figure 11.3: Simulated waveforms for the transition between single-phase control and two-phase interleaved control from two-phase interleaved to single-phase modes.

Figure 11.4: Simulated waveforms for the transition between single-phase control and two-phase interleaved control, single-phase to two-phase interleaved modes.

Case-2 proposed interleaved boost converter with closed loop operation.

Figure 12: Matlab/Simulink model of the proposed single phase converter with closed loop operation.

Figure 12.1: Simulated output wave forms of the closed loop control of the single phase converter with reference value 324V.
4. Conclusion

The contributions of this paper include:
1) Proposal of a new integrated inverter/converter circuit of motor drives with dual-mode control for EV/HEV applications to significantly reduce the volume and weight;
2) Proposal of a new control method for the integrated inverter/converter circuit operating in boost converter mode to increase the efficiency;
3) Verification of the proposed integrated inverter/converter circuit;
4) Verification of the proposed control method. Experimental results show that the voltage boost ratio can go up to 3. Under full-load condition, the maximum efficiency is more than 96% and efficiency can be maintained at more than 91.9% for voltage ratios varies from 1.25 to 3. These results fully confirm the claimed merits of the proposed integrated circuit and control method.

References