

Figure 7: Schematic of Type A Static NOT

In the DML Type-B Static NOT topology, the switching element is an NMOS transistor connected parallel to the Pull-down network. The input to the switching element is a constant low voltage to make it OFF. The only difference when designing DML Type-B Dynamic NOT topology is that the input to the switching element is a clock signal having pre-charge and evaluate phase for dynamic mode.

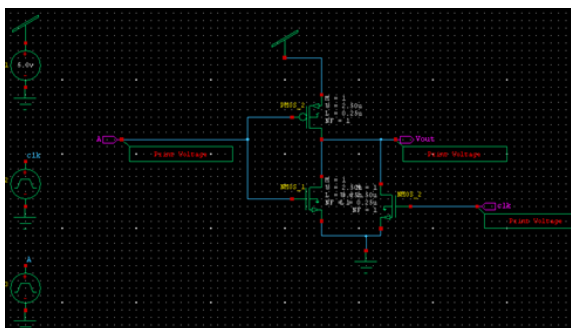


Figure 8: Schematic of Type B Dynamic NOT

The total power consumption in watts for NAND, NOR & NOT logic gates Type A and B is tabulated below.

Table 1: Comparison of NAND, NOR & NOT gates

TYPE	NAND	NOR	NOT
Conventional	6.24E-09W	5.11E-09W	3.35E-09W
Type A Static	5.32E-09W	4.91E-09W	2.37E-09W
Type A Dynamic	9.44E-09W	5.54E-09W	1.15E-09W
Type B Static	6.07E-09W	5.08E-09W	3.27E-09W
Type B Dynamic	7.02E-09W	5.15E-09W	3.26E-09W

4. DML Implementation of a Full Adder

Full Adder is the most used basic circuit in digital electronics. This work has focused on the most commonly used basic CMOS full-adder. Simulations are done for Type A and B Full Adders in Static and Dynamic modes.

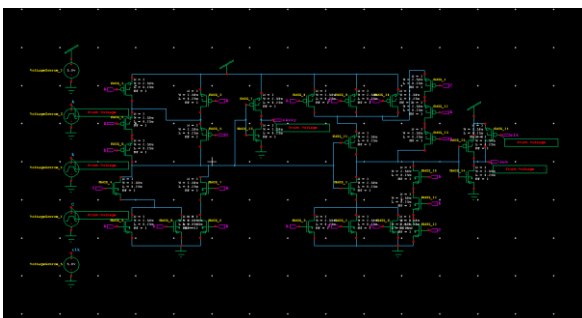


Figure 9: Schematic of Type A Static Full Adder

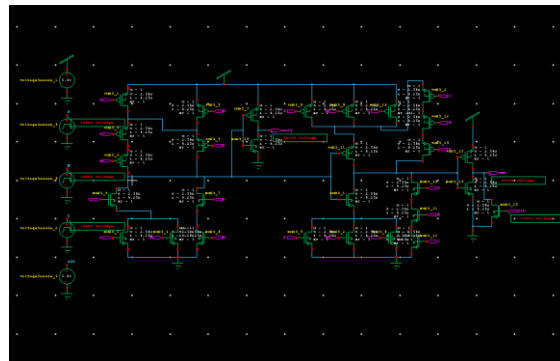


Figure 10: Schematic of Type B Dynamic Full Adder

Table 2: Comparison of Type A & B Full Adder

TYPE	FULL ADDER
CONVENTIONAL	1.978989e-004 watts
TYPE A Static	1.052234e-002 watts
TYPE B Dynamic	3.020891e-003 watts

5. Conclusion

From the results, we can observe that in Static mode there is a power reduction as compared to conventional mode. The DML sizing strategy results in reduced energy dissipation, as compared to conventional static CMOS gates. DML has immunity to process variations, temperature fluctuations, and solving some of the domino's well known drawbacks such as charge sharing, crosstalk noise, and susceptibility to glitches, which intensify with process and voltage scaling. My future extension of this work is the implementation of Power gating techniques so as to further reduce the total power consumption.

References

- [1] Asaf Kaizerman, Sagi Fisher, and Alexander Fish, "Subthreshold Dual Mode Logic," Ieee Transactions On Very Large Scale Integration (Vlsi) Systems, Vol. 21, No. 5, MAY 2013
- [2] M. Alioto, "Ultralow power VLSI circuit design demystified and explained: A tutorial," IEEE Trans. Circuits Syst. I, vol. 59, no. 1, pp.3–29, Jan. 2012.
- [3] A.P.Chandrakasan, S.Sheng and R.W.Brodersen, "Low-power CMOS digital," Solid-State Circuits, IEEE Journal of, vol.27, pp.473-484, 2002.
- [4] D.Bol, R. Ambroise, D. Flandre, and J. D. Legat, "Analysis and minimization of practical energy in 45 nm subthreshold logic circuits," in Proc. IEEE Int. Conf. Comput. Design, Oct. 2008, pp. 294–300.
- [5] N,Verma, J. Kwong and A.P. Chandrakasan, "Nanometer MOSFET variation in minimum energy subthreshold circuits," IEEE Transactions on Electron Devices, vol. 55, pp. 163-174, 2008.
- [6] J. Kao, S. Narendra and A. Chandrakasan, "Subthreshold leakage modeling and reduction techniques," in Proceedings of the 2002 IEEE/ACM international conference on Computer-aided design, pp. 141-148, 2002.
- [7] B. H. Calhoun, A. Wang, and A. Chandrakasan, "Modeling and sizing for minimum energy operation in

- subthreshold circuits,” IEEE J. Solid-State Circuits, vol. 40, no. 9, pp. 1778–1786, Sep. 2005.
- [8] D. Markovic, C. C. Wang, L. P. Alarcon, and J. M. Rabaey, “Ultralowpower design in near-threshold region,” Proc. IEEE, vol. 98, no. 2, pp.237–252, Feb. 2010.
- [9] B. Zhai, S. Hanson, D. Blaauw, and D. Sylvester, “Analysis and mitigation of variability in subthreshold design,” in Proc. Int. Symp. Low Power Electron. Design, Aug. 2005, pp. 20–25.
- [10] R. Swanson and J. Meindl, “Ion-implanted complementary MOS transistors in low-voltage circuits,” IEEE J. Solid-State Circuits, vol. 7, no.2, pp. 146–153, Apr. 1972.
- [11] B. Zhai, L. Nazhandali, J. Olson, A. Reeves, M. Minuth, R. Helfand, S. Pant, D. Blaauw, and T. Austin, “A 2.60 pJ/Inst subthreshold sensor processor for optimal energy efficiency,” in Symp. VLSI Circuits, Dig. Tech. Papers, 2006, pp. 154–155.
- [12] W. M. Pensey and L. Lau, MOS Integrated Circuits. New York: Van Nostrand, 1972, pp. 260–282.
- [13] H. Soeleman, K. Roy, and B. Paul, “Sub-domino logic: Ultralow power dynamic sub-threshold digital logic,” in Proc. 14th Int. Conf. VLSI Design, 2001, pp. 211–214.
- [14] H. Yamada, T. Hotta, T. Nishiyama, F. Murabayashi, T. Yamauchi, and H. Sawamoto, “A 13.3 ns double-precision floating point ALU and multiplier,” in Proc. IEEE Int. Conf. Comput. Design: VLSI Comput. Process., Oct. 1995, pp. 466–470.
- [15] H. Razak, High Performance ASIC Design: Using Synthesizable Domino Logic in an ASIC Flow. Cambridge, U.K.: Cambridge Univ. Press, 2008.

Author Profile



Celine Elsa Jose is currently pursuing her M.E VLSI Design in department of ECE in Hindusthan Institute of Technology, Coimbatore, India. She has completed B.E in Electrical and Electronics Engineering from Karunya University, Coimbatore. Her research areas are VLSI, Low Power Design.



B. Kousalya is an Assistant professor in the Department of Electronics and Communication Engineering, Hindusthan Institute of Technology, Coimbatore, India. She has also worked as Assistant professor at Karpagam College of Engineering, Coimbatore. She received M.E in Applied Electronics from Dr.Mahalingam College of Engineering and technology, Pollachi. Her areas of interest are Low Power VLSI design and Image processing.